

Experimental Characterization of the Electrostatic Levitation Force in MEMS Transducers

Meysam Daeichin

Department of Mechanical Engineering,
Binghamton University,
Binghamton, NY 13902
e-mail: mdaeich1@binghamton.edu

Ronald N. Miles

Department of Mechanical Engineering,
Binghamton University,
Binghamton, NY 13902
e-mail: miles@binghamton.edu

Shahrazad Towfighian¹

Department of Mechanical Engineering,
Binghamton University,
Binghamton, NY 13902
e-mail: stowfigh@binghamton.edu

In this study, a two-step experimental procedure is described to determine the electrostatic levitation force in micro-electromechanical system transducers. In these two steps, the microstructure is excited quasi-statically and dynamically and its response is used to derive the electrostatic force. The experimental results are obtained for a 1 mm by 1 mm plate that employs 112 levitation units. The experimentally obtained force is used in a lumped parameter model to find the microstructure response when it is subjected to different dynamical loads. The natural frequency and the damping ratios in the model are identified from the experimental results. The results show that this procedure can be used as a method to extract the electrostatic force as a function of the microstructure's degrees-of-freedom. The procedure can be easily used for any microstructure with a wide variety of electrode configurations to predict the response of the system to any input excitation. [DOI: 10.1115/1.4046625]

Keywords: dynamics, non-linear vibration, stability, system identification

1 Introduction

The superior performance and successful implementation of capacitive micro-electromechanical system (MEMS) transducers such as MEMS microphones [1], accelerometers [2,3], gyroscopes [4,5], pressure sensors [6], and micro-mirrors [7] in our everyday lives has motivated researchers to study the multi-physics domain of these tiny devices to improve their performance and to address their shortcomings. This improvement will gradually lead to devices for new applications such as endoscopic imaging, metrology, and energy harvesting [8–10].

The most common capacitive transduction scheme in MEMS transducers is to use two electrodes with different voltages that have relative motion with respect to each other [11]. The main drawback of this design is the pull-in instability of the microstructure that happens when the two electrodes get close to each other beyond a threshold gap where the compliant element of the microstructure is no longer able to overcome the electrostatic force [12].

The levitation electrostatic force scheme resolves this issue by omitting the pull-in instability as a microstructure failure mode [13]. In this scheme, there are three fixed electrodes and one moving electrode as shown in Fig. 1(a). The moving electrode and the fixed electrode beneath it have the same voltage (ground). Applying an electrical voltage on the side fixed electrodes creates an electric field that results in an electrostatic force on the moving electrode. This electrostatic force pushes the moving electrode away from the bottom electrode. This is because the bottom surface of the moving electrode is shielded by the bottom electrode, and therefore, there are more lines of electric field going to the top surface of the moving electrode compared to the number of lines going to the bottom surface. As a result, the electrostatic force on the top surface, which is upward, is larger than the downward electrostatic force on the bottom surface. This means the net electrostatic force on the moving electrode is upward, pushing it away from the bottom electrode. This nature of the electrostatic force eliminates the possibility of the moving electrode being pulled into the bottom electrode. It also leads to different types of nonlinear behavior distinct from those for transducers based on parallel plate

configuration which could be beneficial for some applications. For instance, the parametric resonance of transducers based on this scheme can be exploited to operate micromirrors at higher frequencies while maintaining the large displacement for the mirror [14]. It is worth mentioning that the electrostatic force in the levitation scheme, in general, is smaller than the force in the parallel plate scheme because of the configuration of the electrodes. For comprehensive study on the nonlinear dynamics of microstructures based on levitation scheme, one could refer to Refs. [14–16].

In the levitation-based electrode configuration, even if the moving electrode comes into contact with the underlying electrode, there would be no stiction or micro-welding because both electrodes carry the same voltage. This configuration allows for large-amplitude displacement of the moving electrode which is very desirable in applications like micromirrors [14,17,18]. It is worth mentioning that similar performance can be achieved with a slightly different design that employs two fixed electrodes and one moving electrode [19]. These unique characteristics of the levitation scheme have inspired researchers to explore its feasibility for a wide range of applications from micro-mirrors and accelerometers to MEMS

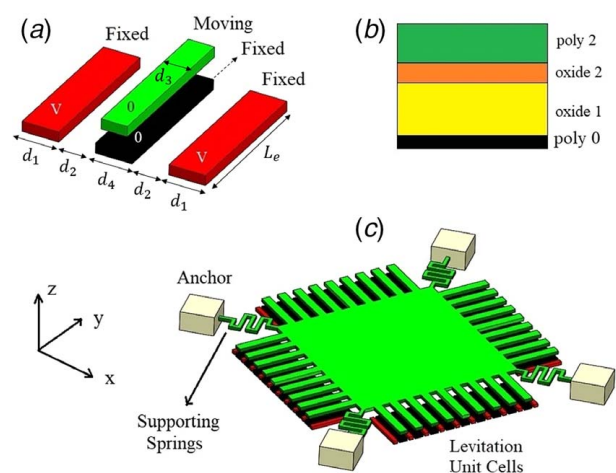


Fig. 1 (a) Repulsive unit cell (not drawn to scale). The dimensions are given in Table 1, (b) schematic of structural and oxide layers used in the fabrication process, and (c) schematic of the microstructure (not drawn to scale).

¹Corresponding author.
Contributed by the Design Engineering Division of ASME for publication in the JOURNAL OF VIBRATION AND ACOUSTICS. Manuscript received July 13, 2019; final manuscript received March 4, 2020; published online March 11, 2020. Assoc. Editor: Slava Krylov.

switches [14,16,20,21]. A comprehensive analysis of the dynamics of the moving electrode in a repulsive electrode configuration is presented in our previous work [15].

To find the electrostatic force on the moving electrode in the levitation scheme, one needs to solve Laplace's equation for the electric field distribution [22–25]. For conventional parallel plate or comb drive configurations, the electric field between the electrodes is commonly assumed to be uniform. This practical assumption leads to a simple analytical solution for Laplace's equation [26] and the electrostatic force. However, finding the electric field in the levitation scheme is more challenging. The electric field in this multi-electrode system is not uniform which makes solving Laplace's equation analytically difficult.

The conformal mapping technique [27] can be used to solve for an analytical solution of the electric field distribution in a multi-electrode system, but using the analytical result in a mathematical model of the microstructure is cumbersome [28]. In the case of an analytical solution being available, the least-squares method can be used to replace the solution with a polynomial form which is mathematically easier to handle.

Finite element and boundary element methods are often used approaches to solve for the electric field in a multi-electrode system which is more common than the analytical approach [20,29]. However, solving the problem in a three-dimensional (3D) domain is computationally difficult and inefficient. To circumvent this complexity, the lengths of the electrodes are usually assumed to be infinite so that the problem could be solved in 2D which compromises the accuracy of the solution [29–31].

Furthermore, these analytical or finite element solutions are usually achieved for just one unit cell. They do not account for multiple unit cells used next to each other when attached to a larger microstructure. For example, Ozdogan et al. have used multiple arch-shape unit cells attached to a circular micro-mirror plate with 1 mm diameter [14]. In another work by Daichin et al. [16], 81 rectangular-shaped unit cells are attached to the 1 mm by 1 mm proof-mass plate of an accelerometer. The existence of fabrication imperfections in the microstructure causes the real unit cells to deviate from the simplified two-dimensional (2D) model. For instance, in the accelerometer with these unit cells in Ref. [16], the proof mass has an initial curvature of 6 μm because of the residual stress created during the fabrication process. This curvature affects how the moving electrodes of the levitation units are initially placed above the fixed electrodes and how they move. The difference between the real-world microstructure and the finite element model leads to an inaccurate prediction of the electrostatic force.

In this study, we have proposed a two-step experimental procedure to obtain the levitation electrostatic force in MEMS transducers. The procedure is explained on one microstructure, but it can be used for any microstructure with arbitrary geometry based on levitation electrostatic force. Characterizing the levitation electrostatic force is important for devices that have large surface to thickness ratios and are susceptible to residual stresses and deformation. For these devices, two-dimensional finite element analysis (FEA) does not yield an accurate representation of the electrostatic force, and three-dimensional FEA is computationally expensive. The effectiveness of the proposed approach is examined using the electrostatic force in a lumped parameter model to predict the frequency response of the microstructure when subjected to a dynamical excitation. The experimentally obtained electrostatic force in this procedure accounts for all the fabrication imperfections and fringing effects of multiple unit cells next to each other for any shape of a levitation unit cell. This procedure is non-invasive and does not lead to any structural or electrical damage to the microstructure.

The working principle and fabrication of the microstructure presented in this study are explained in Sec. 2. In Sec. 3, the experimental procedure and results are presented. In Sec. 4, simulation results with the obtained electrostatic force are compared with experimental results when the microstructure is subjected to different dynamical loads. The conclusion is given in Sec. 5.

2 Working Principle and Fabrication

The schematic of the microstructure considered in this paper is shown in Fig. 1(c). This microstructure consists of a square plate that is anchored by four supporting springs. The levitation unit cells are employed on all four sides of this plate. When a bias voltage is applied on the side electrodes, the moving electrodes of the levitation unit cells are pushed away from the substrate and lift the microstructure plate up. Depending on the application, this mechanism can be used for sensing and actuation.

For sensing, the microstructure plate would be excited with an external force to vibrate and the change in capacitance between the electrodes is measured. For instance, if used as a MEMS microphone [1], the plate is going to vibrate from an incoming sound pressure wave and the capacitance change is measured. Or, for a MEMS accelerometer [2], the microstructure, when subjected to a base excitation, is going to vibrate because of its inertia. This vibration leads to a change in capacitance that can be measured with an electric circuit [32] and related to base acceleration.

The fabricated microstructure is shown in Fig. 2. The fabrication is done using a POLYMUMPS standard process that allows for three structural polysilicon layers with 0.5 μm , 2 μm , and 1.5 μm thicknesses and two oxide layers with 2 μm and 0.75 μm thicknesses from bottom to up, respectively.

In the fabricated device, shown in Fig. 2, the three fixed electrodes are made in the first polysilicon layer (0.5 μm thick which is called poly 0 in the standard) and then the second polysilicon layer (poly 1 in the standard) is skipped. The microstructure and moving electrodes are fabricated in the third structural layer (poly 2 in the standard with a thickness of 1.5 μm). Skipping the second layer allows us to stack the two oxide layers on top of each other which when etched provides a 2.75 μm gap between the microstructure and its underlying substrate (poly 0 layer). The schematic of the layers before etching is shown in Fig. 1(b).

As shown in Fig. 2, there are multiple etch holes equally spaced from each other to ensure the successful releasing of the microstructure during the etching process.

3 Experimental Procedure and Results

This section entails the general procedure to obtain the electrostatic force for the MEMS device shown in Fig. 2. This microstructure consists of a 1 mm by 1 mm perforated plate which is supported by four serpentine springs at its corners. On each side of the plate,

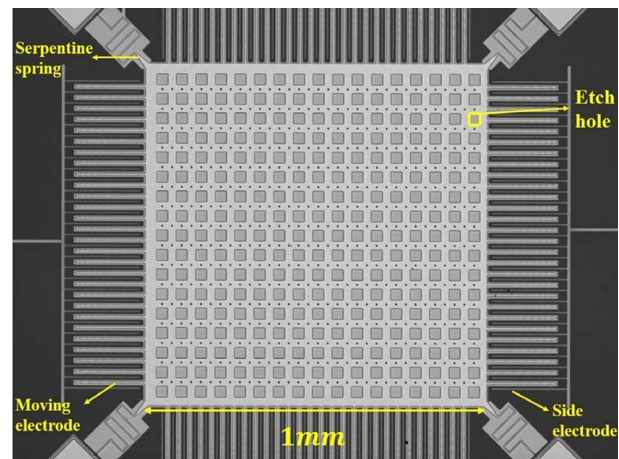


Fig. 2 Top view of the fabricated microstructure. The moving and side electrodes are shown in this figure. The bottom electrodes are underneath the moving electrodes. The microstructure is supported with four serpentine springs. There are several etch holes on the microstructure plate to ensure the release of the microstructure.

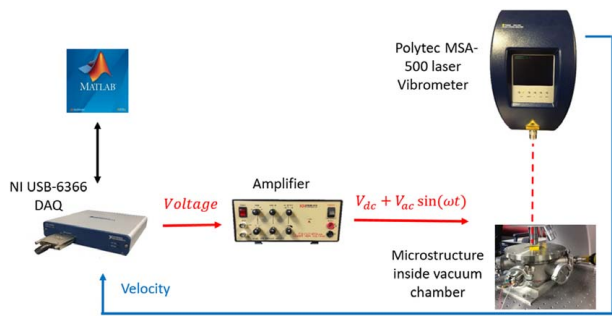


Fig. 3 Experimental setup. The AC voltage is sent from the NI USB-6366 DAQ to the amplifier. The amplifier amplifies this signal and adds the desired DC component to it and then sends it to the microstructure inside the vacuum chamber. The motion of the microstructure is measured with the polytech MSA-500 laser vibrometer by reading the velocity.

there are 28 sets of levitation unit cells. The degree-of-freedom (DOF) is considered to be translational in the out-of-plane direction (z direction). The goal is to find the electrostatic force on the microstructure as a function of its DOF. There are two measurements that need to be done to collect all the data necessary to calculate the electrostatic force normalized with respect to the effective mass of the microstructure. The experimental setup is shown in Fig. 3.

The first measurement is obtained by applying a quasi-static voltage on the side electrodes and recording the displacement with the Polytec MSA-500 laser vibrometer and NI USB 6366 DAQ. A sine function with a duration several orders of magnitudes larger than the microstructure's natural period is chosen for the voltage profile. This large duration of applied load ensures that the microstructure response is quasi-static and therefore the dynamical effects such as viscous energy dissipation and acceleration, which are proportional to the first and second time derivatives of displacement, are negligible. The continuity of the voltage profile and its derivatives with respect to time avoids exerting shock-like excitation on the microstructure. This minimizes the effect of damping or dynamical effects in the displacement of the microstructure. Figure 4 shows the applied voltage and the displacement of the microstructure versus time. Higher voltages lead to larger displacements that result in measuring the electrostatic force for a larger range of DOF. For instance, consider a case where we want to measure the electrostatic force when the microstructure is moving

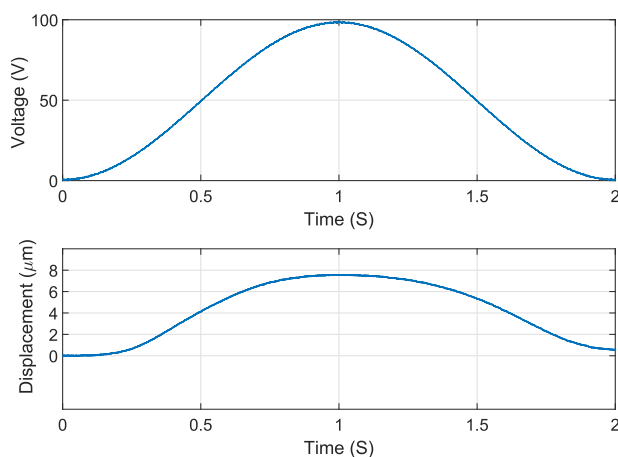


Fig. 4 By applying a quasi-static sinusoidal voltage on the side electrodes, the microstructure undergoes a quasi-static displacement with a similar profile. In this measurement, the displacement of the microstructure is recorded with a vibrometer displacement decoder. At the end of the measurement, the displacement is not exactly zero which is because of the drift in the vibrometer reading.

from its initial rest position to $5 \mu\text{m}$ above this position. In order to have the electrostatic force in this range, we have to apply enough voltage during the experimental procedure to have the microstructure move $5 \mu\text{m}$. This means that if we apply higher voltages during the experimental procedure, we can then calculate the electrostatic force for larger range of microstructure displacement. In our calculation, we have considered the first half of the voltage and displacement data where the voltage and displacement go from 0 to their maximum values. According to Fig. 4, the displacement at the end of the quasi-static loading (when voltage goes zero) has a small nonzero value. This is because of the inherent drift in laser vibrometer measurement of displacement. If the load time duration is large enough that the drift of the laser over time is significant, then this drift should be removed using a least-square method. Here, we have ignored this drift because it is small, especially in the first part of the displacement data where displacement goes from 0 to its maximum value (not from maximum to 0). Because this experiment is quasi static, the pressure is not important so we have conducted this measurement at atmospheric pressure.

The second measurement is designed to obtain the linearized natural frequency (the resonance frequency at small excitation levels) of the microstructure at different DC voltages on side electrodes. As explained in Refs. [14–16], the resonance frequency of a microstructure that employs levitation unit cells increases with the DC voltage on the side electrodes. To measure the linearized natural frequency of the microstructure at a desired DC voltage, first, the DC voltage is applied on the side electrodes. Then, the microstructure is excited slightly while the velocity at one point (here, at the center of the plate) is being recorded with the laser vibrometer. There are three ways to excite the microstructure to get the linearized frequency data.

The first way to excite the microstructure is to apply a small random voltage on the side electrodes (after a DC voltage is initially applied on side electrodes). The frequency at which the magnitude of the transfer function between the random voltage (input) and velocity (output) is maximum yields the linearized natural frequency. The second way is to apply a small AC voltage on top of the initial DC voltage and perform a frequency sweep while measuring the velocity of the device. The peak in the frequency response again yields the linearized natural frequency at the given applied DC voltage. The third way to excite the microstructure is to excite it mechanically with a speaker or a shaker and performing a frequency sweep to obtain the linearized natural frequency. In this study, we have chosen the second way to extract the linearized natural frequencies. It is worth mentioning that all these techniques lead to the same results but the important consideration here is to perturb the microstructure slightly when a DC voltage is applied on the side electrodes. The reason behind this is that the levitation electrostatic units exhibit different nonlinear behaviors like softening and hardening when they undergo large deflections [14–16], or parametric and superharmonic resonances when they are excited at half or twice of their fundamental natural frequency [14,33]. In Sec. 4, we will use these experimentally obtained resonance frequencies to equate them with linearized natural frequencies obtained from stability analysis using the Jacobian matrix. Linearized natural frequencies represent the microstructure resonance frequencies around their equilibrium state when it is perturbed slightly and this is why it is important to apply small excitation and not provoke any nonlinearities that accompany large excitations and displacements.

It is important to keep track of the pressure in the resonance frequency experiment because the air might introduce stiffness effects to the microstructure. This is because of the squeeze film effect that creates a stiffness and an energy dissipation mode that affects the microstructure dynamics. A comprehensive study of the squeeze film damping effect is given in Ref. [26]. It is recommended to perform this test in pressure ranges in which the microstructure is going to be used. Here, we have used reduced pressure range of 277–287 (mTorr) for the natural frequency and all the dynamical experiments. The measured linearized natural frequency at different voltages is shown in Fig. 5. It should be mentioned that, if needed,

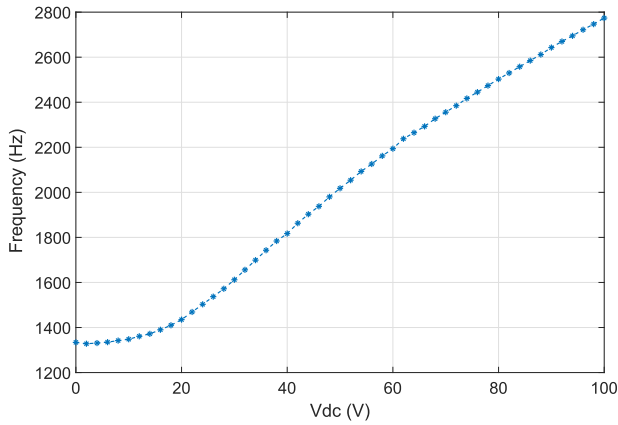


Fig. 5 Resonance frequency for different voltages on the side electrodes is measured. By increasing the voltage on the side electrode, the electrical stiffness increases which lead to an increase in the resonance frequency of the microstructure.

the effective mass of the microstructure can be calculated by measuring the effective stiffness of the microstructure and using the fundamental natural frequency. The effective stiffness of the microstructure can be measured using a nano-indenter to apply a mechanical force and record the microstructure's displacement when there is no voltage on the side electrodes. This procedure is explained in more detail in Ref. [34]. Here, because we will use a lumped parameter model that is normalized with respect to effective mass (Sec. 4), there is no need for measuring the effective mass.

4 Mathematical Modeling

A lumped parameter model with one degree-of-freedom is considered in Eq. (1) to simulate the microstructure's response at the center of the plate (x).

$$\ddot{x} + 2\zeta\omega_n\dot{x} + \omega_n^2x = f(x, V) \quad (1)$$

where $f(x, V)$ is the electrostatic force normalized with respect to the effective mass of the microstructure, ζ is the damping ratio, and ω_n is the mechanical natural frequency of the microstructure.

4.1 Quasi-Static Motion. As mentioned before, the quasi-static experiment is done in a way that dynamical effects like acceleration and damping would be negligible. Therefore, by putting the time derivative terms in Eq. (1) equal to zero, the equation governing the static response (x_{st}) of the microstructure is obtained (Eq. (2))

$$\omega_n^2x_{st} = f(x_{st}, V_{dc}) \quad (2)$$

This means that as ω_n is known from the resonance frequency measurement, the normalized electrostatic force is known at any equilibrium point (x_{st}, V_{dc}).

It is worth mentioning that $f(x, V)$ in Eq. (1) is a scalar function of two variables (x and V), and therefore, we can define the gradient operator acting on it to calculate its directional derivative along any arbitrary direction in the x - V plane. Using this concept, we can calculate the directional derivative of $f(x, V)$ at each equilibrium point along the vector (\vec{u}) in the x - V plane that is tangent to the x_{st} - V_{dc} curve as given in Eq. (3). Figure 6 is useful to visualize this directional derivative. Because we have x_{st} - V_{dc} from the experiment (blue curve in Fig. 6), we can find the tangent line to it at each point.

$$df(x, V)_{\vec{u}} = \vec{\nabla}f(x, V) \cdot \vec{u} = \frac{\partial f(x_{st}, V_{dc})}{\partial x}u_1 + \frac{\partial f(x_{st}, V_{dc})}{\partial V}u_2 \quad (3)$$

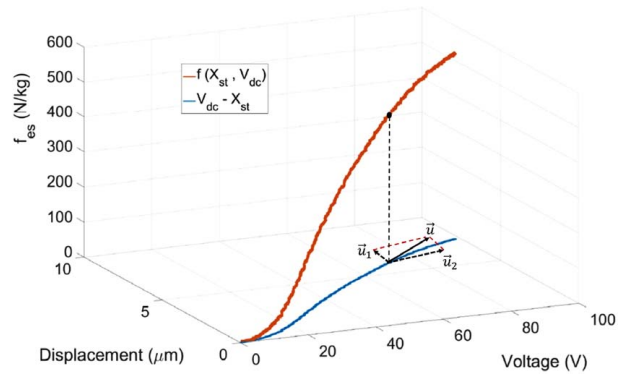


Fig. 6 Electrostatic force and displacement are shown as the voltage on the side electrode varies. The gradient of the electrostatic force is calculated along a unit vector tangent to the voltage–displacement curve in the X - V plane. (Color version online.)

where $\vec{\nabla}$ is the gradient operator and

$$\vec{u} = u_1\vec{i}_x + u_2\vec{j}_V \quad (4)$$

is the unit vector tangent to the x_{st} - V_{dc} curve at (x_{st}, V_{dc}). Unit vectors along x and V axes are \vec{i}_x and \vec{j}_V , respectively.

According to Eq. (2), the directional derivative of $f(x, V)$ along the (x_{st}, V_{dc}) is

$$df(x, V)_{\vec{u}} = \omega_n^2\Delta x_{st} \quad (5)$$

Therefore, Eqs. (3) and (5) give us

$$\frac{\partial f(x_{st}, V_{dc})}{\partial x}u_1 + \frac{\partial f(x_{st}, V_{dc})}{\partial V}u_2 = \omega_n^2\Delta x_{st} \quad (6)$$

4.2 Resonance Frequency. To find the linearized natural frequencies from the model given in Eq. (1), first, we need to write the governing equation in state space. By using state variables x_1 and x_2 for displacement (x) and velocity (\dot{x}), respectively, the governing equation of motion can be written in state space as described as follows:

$$\begin{cases} \dot{x}_1 = x_2 = g_1(x_1, x_2) \\ \dot{x}_2 = -2\zeta\omega_nx_2 - \omega_n^2x_1 + f(x_1, V) = g_2(x_1, x_2) \end{cases} \quad (7)$$

Then, the Jacobian matrix can be constructed as illustrated as follows:

$$J = \begin{bmatrix} \frac{\partial g_1}{\partial x_1} & \frac{\partial g_1}{\partial x_2} \\ \frac{\partial g_2}{\partial x_1} & \frac{\partial g_2}{\partial x_2} \end{bmatrix} \quad (8)$$

The eigenvalues of the Jacobian matrix for the undamped system yield the linearized natural frequencies at the equilibrium points ($x_1 = x_{st}$ and $x_2 = 0$). Equation (9) gives the eigenvalue problem

$$|J - \lambda I| = \begin{vmatrix} -\lambda & 1 \\ -\omega_n^2x_{st} + \frac{\partial f(x_{st}, V_{dc})}{\partial x} & -\lambda \end{vmatrix} = 0 \quad (9)$$

On the other hand, the eigenvalue of the Jacobian matrix (λ) is related to the linearized natural frequency through a complex equation given in Eq. (10):

$$\lambda = i\omega \quad (10)$$

where ω is the linearized natural frequency. Solving Eq. (9) for λ using Eq. (10) and rearranging the terms gives

$$\frac{\partial f(x_{st}, V_{dc})}{\partial x} = \omega_n^2 - \omega^2 \quad (11)$$

Equations (6) and (11) construct a linear system of algebraic equations with two equations and two unknowns, $\partial f(x_{st}, V_{dc})/\partial x$ and $\partial f(x_{st}, V_{dc})/\partial V$, that can be easily solved to find the unknowns. After finding these two unknowns, the normalized electrostatic force can be obtained using a first-order Taylor expansion for a two-variable function (Eq. (12))

$$f(x, V) = f(x_{st}, V_{dc}) + \frac{\partial f_{es}(x_{st}, V_{dc})}{\partial x}(x - x_{st}) + \frac{\partial f_{es}(x_{st}, V_{dc})}{\partial V}(V - V_{dc}) + \dots \quad (12)$$

This force can be used in the lumped parameter model to predict the microstructure response to dynamical loads. The equation for the lumped parameter model when a DC+AC voltage load is derived using Eqs. (1) and (12).

$$\ddot{x} + 2\zeta\omega_n\dot{x} + \omega_n^2x = f(x_{st}, V_{dc}) + \frac{\partial f(x_{st}, V_{dc})}{\partial x}(x - x_{st}) + \frac{\partial f(x_{st}, V_{dc})}{\partial V}(V - V_{dc}) + \dots \quad (13)$$

The DC + AC voltage that is applied to the side electrodes can be written as

$$V = V_{dc} + V_{ac} \sin(\omega_e t) \quad (14)$$

where V_{dc} and V_{ac} are the DC and AC voltage amplitudes, respectively, and ω_e is the excitation frequency. The microstructure response to this load can be split into two parts, a static part (x_{st}) which is caused by V_{dc} and a dynamic part (x_d) which is from $V_{ac} \sin(\omega_e t)$, i.e.,

$$x = x_{st} + x_d \quad (15)$$

where x_{st} is picked from the quasi-static experiment for the corresponding V_{dc} . Substituting this change of variable into Eq. (13) yields

$$\ddot{x}_d + 2\zeta\omega_n\dot{x}_d + \omega_n^2(x_{st} + x_d) = f(x_{st}, V_{dc}) + \frac{\partial f(x_{st}, V_{dc})}{\partial x}(x_{st} + x_d - x_{st}) + \frac{\partial f(x_{st}, V_{dc})}{\partial V} \times (V_{dc} + V_{ac} \sin(\omega_e Tt) - V_{dc}) + \dots \quad (16)$$

Table 1 Parameters

Parameter	Symbol	Value	Unit
Proof mass length	L	1000	μm
Proof mass width	L	1000	μm
Electrode length	L_e	200	μm
Total number of unit cells	N	112	
Voltaged fixed electrode width	d_1	4	μm
Gap between fixed electrodes	d_2	6	μm
Moving electrode width	d_3	6	μm
Ground fixed electrode width	d_4	16	μm
Electrodes and proof thickness	h_1	1.5	μm
Initial gap	h_2	2.75	μm
Mechanical natural frequency	ω_n	$2\pi \times 1334$	rad/S
Damping ratio ($V_{dc} = 40$)	ξ_0	0.064	
Damping ratio ($V_{dc} = 50$)	ξ_0	0.071	
Damping ratio ($V_{dc} = 60$)	ξ_0	0.080	

Using Eq. (2) and some mathematical simplification and assuming that the higher-order terms in the Taylor series expansion are small for small AC excitations, Eq. (16) can be rewritten in the following form:

$$\ddot{x}_d + 2\zeta\omega_n\dot{x}_d + \omega_n^2x_d = \frac{\partial f(x_{st}, V_{dc})}{\partial x}x_d + \frac{\partial f(x_{st}, V_{dc})}{\partial V}V_{ac} \sin(\omega_e Tt) \quad (17)$$

The coefficients on the right-hand side are now identified, and Eq. (17) can be solved for different load cases (V_{dc} and V_{ac}) to find the dynamical displacement of the microstructure (x_d). All the other parameters in this equation are known except for the damping ratio which is identified from comparing the resonance peaks at different DC voltages given in Table 1. The dependency of the damping ratio to the voltage on side electrodes is explained in Refs. [14,16]. For a comprehensive review of quality factor tuning in MEMS resonators, one can refer to Ref. [35]. Equation (17) can be solved using the shooting method explained in Ref. [26]. All the simulated frequency responses in the paper are obtained by solving this equation using the shooting method.

Because the first-order Taylor expansion is used to obtain the electrostatic force, the force and the model should give better results for small excitations compared to large excitations. Figures 7–12 show the lumped model results compared to experimental results for different voltage excitation. A summary of the results is given in Table 2. In this table, the bandwidth is considered to be the width of the frequency range where the dynamical amplitude is larger than $\frac{1}{\sqrt{2}}$ of the resonance amplitude peak. Comparing Figs. 7, 9, and 11 reveals the increase in the resonance frequency by an increase in the voltage on the side electrodes. For example, the

Table 2 Comparing experimental and simulation results for different load cases

Load case	Peak frequency (Hz)			Peak amplitude (μm)			Bandwidth (Hz)		
	Exp	Sim	Error (%)	Exp	Sim	Error (%)	Exp	Sim	Error (%)
$V_{dc} = 40 - V_{ac} = 1$	1780	1790	0.56	1.08	1.08	0	150	173	15
$V_{dc} = 40 - V_{ac} = 2$	1776	1792	0.9	2.16	2.16	0	148	170	8
$V_{dc} = 40 - V_{ac} = 3$	1764	1792	1.59	3.25	3.24	-0.28	147	170	16
$V_{dc} = 40 - V_{ac} = 4$	1752	1790	2.2	4.26	4.33	1.5	156	170	9
$V_{dc} = 50 - V_{ac} = 1$	1982	1990	0.4	0.97	0.97	0	144	190	32
$V_{dc} = 50 - V_{ac} = 2$	1976	1990	0.7	1.94	1.95	0.5	146	193	32
$V_{dc} = 50 - V_{ac} = 3$	1968	1990	1.12	2.92	2.92	0	145	191	32
$V_{dc} = 50 - V_{ac} = 4$	1954	1990	1.8	3.92	3.89	0.8	144	191	33
$V_{dc} = 60 - V_{ac} = 1$	2168	2176	0.37	0.85	0.85	0	144	216	50
$V_{dc} = 60 - V_{ac} = 2$	2162	2176	0.65	1.74	1.71	-1.7	139	212	52
$V_{dc} = 60 - V_{ac} = 3$	2156	2176	0.93	2.55	2.56	0.4	142	213	50
$V_{dc} = 60 - V_{ac} = 4$	2146	2176	1.4	3.40	3.42	0.6	142	212	49

Note: The peak amplitudes are compared to the one hundredth of a micrometer.

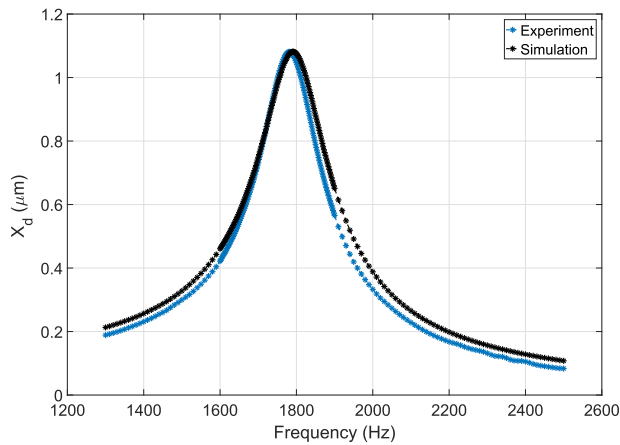


Fig. 7 Dynamical displacement when a time-varying voltage with different frequencies is applied on the side electrodes ($V_{dc} = 40$ (V), $V_{ac} = 1$ (V))

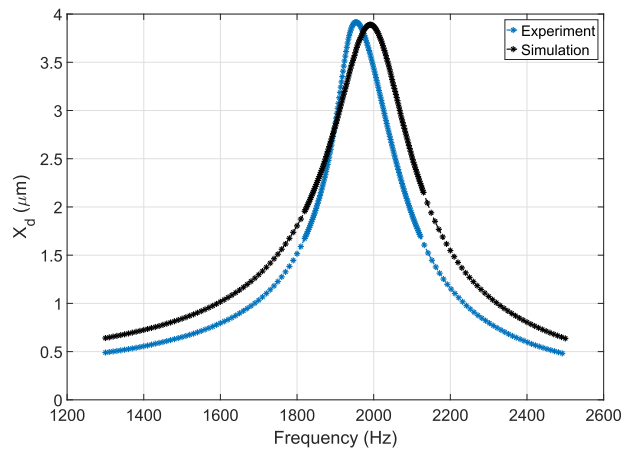


Fig. 10 Dynamical displacement when a time-varying voltage with different frequencies is applied to the side electrodes ($V_{dc} = 50$ (V), $V_{ac} = 4$ (V))

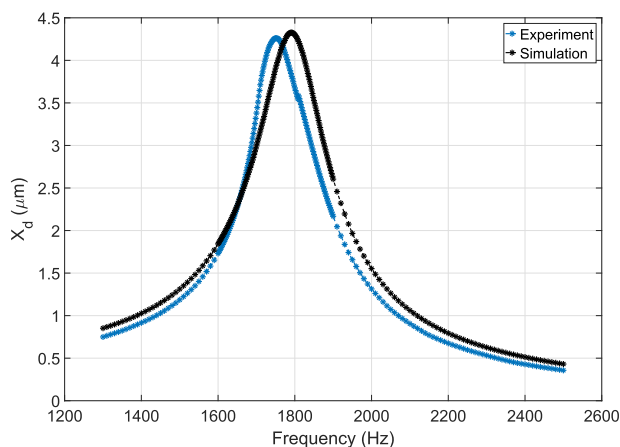


Fig. 8 Dynamical displacement when a time-varying voltage with different frequencies is applied on the side electrodes ($V_{dc} = 40$ (V), $V_{ac} = 4$ (V))

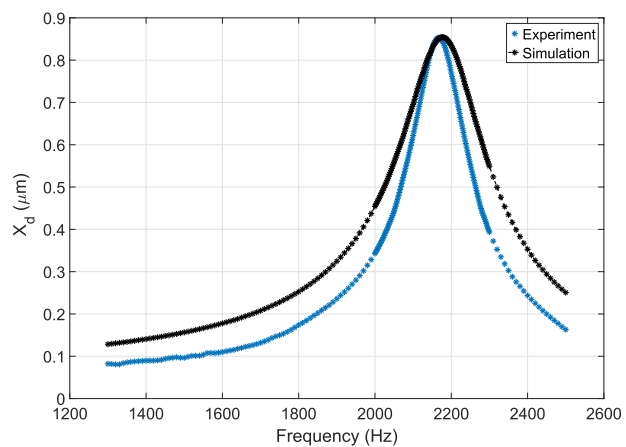


Fig. 11 Dynamical displacement when a time-varying voltage with different frequencies is applied to the side electrodes ($V_{dc} = 60$ (V), $V_{ac} = 1$ (V))

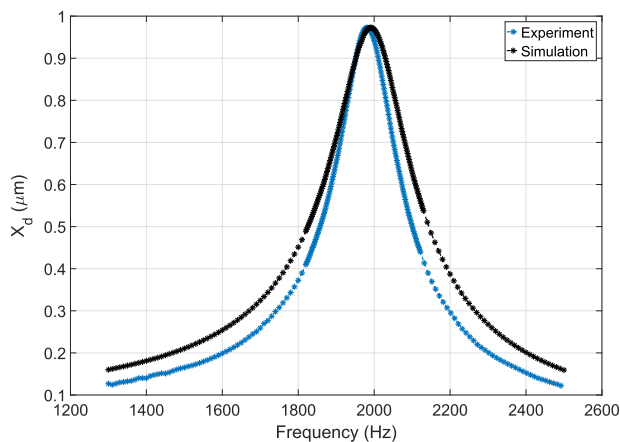


Fig. 9 Dynamical displacement when a time-varying voltage with different frequencies is applied to the side electrodes ($V_{dc} = 50$ (V), $V_{ac} = 1$ (V))

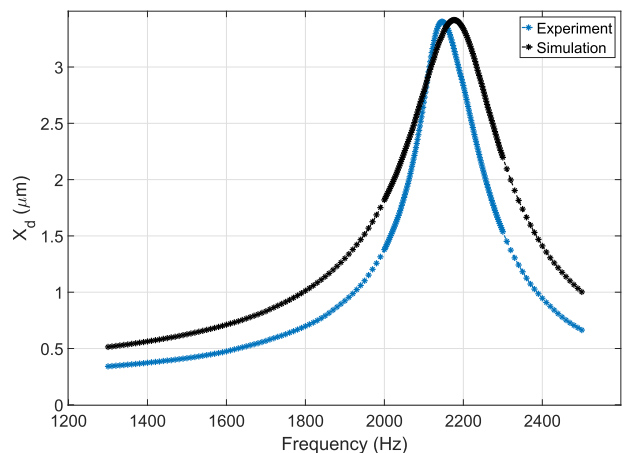


Fig. 12 Dynamical displacement when a time-varying voltage with different frequencies is applied to the side electrodes ($V_{dc} = 60$ (V), $V_{ac} = 4$ (V))

increase of DC excitation voltage to 50(V) from 40(V) shifts the resonant frequency to 1990(Hz) from 1790(Hz). An increase of AC voltage at the same DC level causes an increase of dynamic amplitude (compare Figs. 7 and 8, Figs. 9 and 10, and Figs. 11 and 12). However, an increase of DC voltage at the same AC

amplitude leads to a decrease of dynamic amplitude (Figs. 7, 9, 11 and Figs. 8, 10, 12) because of the increase of electrical stiffness as depicted in Fig. 5. These two properties of levitation transducers are in contrast to parallel-plate actuators that show decrease of electrical stiffness as the DC voltage increases.

It is worth mentioning that at $V_{dc} = 40$, the static displacement of the microstructure is about $2 \mu\text{m}$. This displacement will add to the initial gap between microstructure and the substrate ($2.75 \mu\text{m}$) to provide $4.75 \mu\text{m}$ of room in total for the vibration of the microstructure. Figure 8 shows that the maximum amplitude of the microstructure at its resonance peak is about $4.3 \mu\text{m}$, which means the microstructure is vibrating close to its limit of hitting the substrate. Therefore, even though we obtained the electrostatic force with the small excitation assumption, we can see that the experimentally obtained electrostatic force delivers a very good approximation for when the dynamical amplitude of the microstructure is more than twice as of the static displacement, and the microstructure is vibrating close to its physical limits. However, the slight softening observed in the experimental data could not be captured by the model because of the linear expansion for the electrostatic force (first-order Taylor expansion).

5 Conclusion

In this study, an experimental procedure to obtain the electrostatic levitation force in MEMS transducers with levitation unit cell configurations is presented. The accuracy of the obtained electrostatic force is examined by using it in a lumped parameter model and solving this model for the frequency response of the microstructure to different dynamical loads. A microstructure is fabricated using the POLYMUMPs standard and it is tested with the aid of a laser vibrometer. The simulation and experimental results are compared with each other which shows a close agreement. The effectiveness of the model to capture the dynamical response of the microstructure demonstrates the effectiveness of the proposed procedure to identify the electrostatic force. The proposed method can be employed for non-traditional electrode systems, whose response can deviate from analytical or numerical approaches because of fabrication imperfections and tolerances. The identified electrostatic force from two simplified tests can be used to study the performance of microstructures such as sensors or actuators.

Acknowledgment

The authors would like to acknowledge the financial support of this study by the National Science Foundation (NSF) through grant ECCS 1608692.

References

- [1] Miles, R. N., Cui, W., Su, Q. T., and Homentcovschi, D., 2015, "A MEMS Low-Noise Sound Pressure Gradient Microphone With Capacitive Sensing," *J. Microelectromech. Syst.*, **24**(1), pp. 241–248.
- [2] Guney, M. G., Li, X., Chung, V. P., Paramesh, J., Mukherjee, T., and Fedder, G. K., 2018, "High Dynamic Range CMOS-MEMS Capacitive Accelerometer Array," 2018 IEEE Micro Electro Mechanical Systems (MEMS), Belfast, Northern Ireland, UK, Jan. 2018, pp. 992–995.
- [3] Mukhiya, R., Agarwal, P., Badjatya, S., Garg, M., Gaikwad, P., Sinha, S., Singh, A. K., and Gopal, R., 2019, "Design, Modelling and System Level Simulations of DRIE-Based MEMS Differential Capacitive Accelerometer," *Microsyst. Technol.*, **25**(9), pp. 3521–3532.
- [4] Yang, C., Tang, S., and Tavassolian, N., 2017, "Utilizing Gyroscopes Towards the Automatic Annotation of Seismocardiograms," *IEEE Sens. J.*, **17**(7), pp. 2129–2136.
- [5] Sheikhalah, A., Jafari, K., and Abedi, K., 2019, "Design and Analysis of a Novel MOEMS Gyroscope Using an Electrostatic Comb-Drive Actuator and an Optical Sensing System," *IEEE Sens. J.*, **19**(1), pp. 144–150.
- [6] Hasan, M. H., Alsalem, F. M., and Ouakad, H. M., 2018, "Novel Threshold Pressure Sensors Based on Nonlinear Dynamics of MEMS Resonators," *J. Micromech. Microeng.*, **28**(6), p. 065007.
- [7] Nabholz, U., Heinzlmann, W., Mehner, J. E., and Degenfeld-Schonburg, P., 2018, "Amplitude- and Gas Pressure-Dependent Nonlinear Damping of High-Q Oscillatory MEMS Micro Mirrors," *J. Microelectromech. Syst.*, **27**(3), pp. 383–391.
- [8] Li, H., Duan, X., and Wang, T. D., 2017, "An Electrostatic MEMS Scanner With In-Plane and Out-of-Plane Two-dimensional Scanning Capability for Confocal Endoscopic in Vivo Imaging," 2017 IEEE 30th International Conference on Micro Electro Mechanical Systems (MEMS), Las Vegas, NV, Jan. 2017, pp. 514–517.

- [9] Stange, A., Imboden, M., Javor, J., Barrett, L. K., and Bishop, D. J., 2019, "Building a Casimir Metrology Platform With a Commercial MEMS Sensor," *Microsyst. Nanoeng.*, **5**(1), pp. 1–9.
- [10] Saadatnia, Z., Esmailzadeh, E., and Naguib, H. E., 2019, "High Performance Triboelectric Nanogenerator by Hot Embossing on Self-Assembled Micro-Particles," *Adv. Eng. Mater.*, **21**(1), p. 1700957.
- [11] Mohammed, Z., Elfadel, I. M., and Rasras, M., 2018, "High Dynamic Range z-Axis Hybrid Spring MEMS Capacitive Accelerometer," 2018 Symposium on Design, Test, Integration Packaging of MEMS and MOEMS (DTIP), Roma, Italy, May 2018, pp. 1–4.
- [12] Nayfeh, A. H., Younis, M. I., and Abdel-Rahman, E. M., 2007, "Dynamic Pull-In Phenomenon in MEMS Resonators," *Nonlinear Dyn.*, **48**(1–2), pp. 153–163.
- [13] Wang, W., Ren, H., Ma, W., Qiu, C., Chen, Z., and Fan, B., 2016, "Electrostatic Repulsive Out-of-Plane Actuator Using Conductive Substrate," *Sci. Rep.*, **6**(35118).
- [14] Ozdogan, M., Daeichin, M., Ramini, A., and Towfighian, S., 2017, "Parametric Resonance of a Repulsive Force MEMS Electrostatic Mirror," *Sens. Actuators. A*, **265**, pp. 20–31.
- [15] Pallay, M., Daeichin, M., and Towfighian, S., 2017, "Dynamic Behavior of an Electrostatic MEMS Resonator With Repulsive Actuation," *Nonlinear Dyn.*, **89**(2), pp. 1525–1538.
- [16] Daeichin, M., Ozdogan, M., Towfighian, S., and Miles, R. N., 2019, "Dynamic Response of a Tunable MEMS Accelerometer Based on Repulsive Force," *Sens. Actuators. A*, **289**, pp. 34–43.
- [17] Li, H., Barnes, P., Harding, E., Duan, X., Wang, T. D., and Oldham, K. R., 2019, "Large-Displacement Vertical Electrostatic Microactuator Dynamics Using Duty-Cycled Softening/Stiffening Parametric Resonance," *J. Microelectromech. Syst.*, **28**(3), pp. 351–361.
- [18] Towfighian, S., Selem, A., Abdel-Rahman, E., and Hepler, G., 2011, "A Large-Stroke Electrostatic Micro-Actuator," *J. Micromech. Microeng.*, **21**(7), p. 075023.
- [19] Lee, K. B., and Cho, Y. H., 2001, "Laterally Driven Electrostatic Repulsive-Force Microactuators Using Asymmetric Field Distribution," *J. Microelectromech. Syst.*, **10**(1), pp. 128–136.
- [20] Pallay, M., and Towfighian, S., 2018, "A Reliable MEMS Switch Using Electrostatic Levitation," *Appl. Phys. Lett.*, **113**(21), p. 213102.
- [21] Pallay, M., Ibrahim, A. I., Miles, R. N., and Towfighian, S., 2019, "Pairing Electrostatic Levitation With Triboelectric Transduction for High-Performance Self-Powered MEMS Sensors and Actuators," *Appl. Phys. Lett.*, **115**(13), p. 133503.
- [22] Ouakad, H. M., 2013, "Structural Behavior of Microbeams Actuated by Out-of-Plane Electrostatic Fringing-Fields," Proceedings of the ASME 2013 International Mechanical Engineering Congress and Exposition, San Diego, CA, Nov. 2013.
- [23] Ouakad, H. M., 2015, "Numerical Model for the Calculation of the Electrostatic Force in Non-Parallel Electrodes for MEMS Applications," *J. Electrostat.*, **76**, pp. 254–261.
- [24] Ouakad, H. M., and Sedighi, H. M., 2019, "Static Response and Free Vibration of MEMS Arches Assuming Out-of-Plane Actuation Pattern," *Int. J. Non Linear Mech.*, **110**, pp. 44–57.
- [25] Ouakad, H. M., and Najjar, F., 2019, "Nonlinear Dynamics of MEMS Arches Assuming Out-of-Plane Actuation Arrangement," *ASME J. Vib. Acoust.*, **141**(4), p. 041010.
- [26] Younis, M. I., 2008, *MEMS Linear and Nonlinear Statics and Dynamics*, Wiley, New York.
- [27] Hah, D., 2018, "Analytical Design of Linear Variable Capacitors With Shaped-Finger Comb-Drive Actuators," 2018 Symposium on Design, Test, Integration Packaging of MEMS and MOEMS (DTIP), Roma, Italy, May 2018, pp. 1–5.
- [28] He, S., and BenMrad, R., 2005, "Large-Stroke Microelectrostatic Actuators for Vertical Translation of Micromirrors Used in Adaptive Optics," *IEEE Trans. Ind. Electron.*, **52**(4), pp. 974–983.
- [29] Krylov, S., Ilic, B. R., and Lulinsky, S., 2011, "Bistability of Curved Microbeams Actuated by Fringing Electrostatic Fields," *Nonlinear Dyn.*, **66**(3), pp. 403–426.
- [30] Krylov, S., Molinazzi, N., Shmilovich, T., Pomerantz, U., and Lulinsky, S., 2010, "Parametric Excitation of Flexural Vibrations of Micro Beams by Fringing Electrostatic Fields," Proceedings of the ASME 2010 International Design Engineering Technical Conferences and Computers and Information in Engineering Conference, IDETC-CIE 2010, Montreal, QC, Aug. 2010.
- [31] Linzon, Y., Ilic, B., Lulinsky, S., and Krylov, S., 2013, "Efficient Parametric Excitation of Silicon-on-Insulator Microcantilever Beams by Fringing Electrostatic Fields," *J. Appl. Phys.*, **113**(16), p. 163508.
- [32] Miles, R. N., 2018, "A Compliant Capacitive Sensor for Acoustics: Avoiding Electrostatic Forces at High Bias Voltages," *IEEE Sens. J.*, **18**(14), pp. 5691–5698.
- [33] Botello, M., Beatriz, J., and Caruntu, D. I., 2018, "Voltage Response of Circular Plate MEMS Resonators Under Superharmonic Resonance," ASME International Mechanical Engineering Congress and Exposition, 4B: Dynamics, Vibration, and Control, Pittsburgh, PA, Nov. 2018.
- [34] Younis, M. I., Alsalem, F. M., Miles, R. N., and Su, Q., 2007, "Characterization of the Performance of Capacitive Switches Activated by Mechanical Shock," *J. Micromech. Microeng.*, **17**(7), p. 1360.
- [35] Miller, J. M. L., Ansari, A., Heinz, D. B., Chen, Y., Flader, I. B., Shin, D. D., Villanueva, L. G., and Kenny, T. W., 2018, "Effective Quality Factor Tuning Mechanisms in Micromechanical Resonators," *Appl. Phys. Rev.*, **5**(4), p. 041307.